A OUR HARDWARE APPROACH

This section presents the *Schrödinger's FP* hardware encoder/decoder units that efficiently exploit the potential created by our quantization schemes. Without the loss of generality, we describe compressors/decompressors that process groups of 8 FP32 values. This section assumes that the reader is aware of prior work that demonstrated why it is possible and desirable to encode tensor values using variable length containers when storing them to external DRAM, e.g., (Han et al., 2016c; Judd et al., 2016a; Han et al., 2016a).

Overview of Hardware: At high-level, our hardware compressors/decompressors transparently encode/decode tensor values just before the memory controller. When values are stored to external DRAM, the encoders efficiently encode the values to use as few bits as necessary. When values are read back from external DRAM, the decoders, expand the values to the original format. This way the rest of the on-chip memory hierarchy and compute units can remain as-is.

Compressor: The compressor accepts one row of 8 numbers per cycle. In the compressor's first stage, it subtracts the fixed bias from the exponents. The resulting differences along with mantissas are then processed by 8 Packer units and a width detector as shown in Figure 9. The mantissa quantizer method, whether Quantum Mantissa or BitWave, provides the same mantissa length for all values. Each value within the row is encoded using the same number of bits, calculated as the sum of the provided mantissa bitlength and the bitlength needed to store the largest exponent difference across the row. The width detector, as its name suggests, will detect how many bits are needed to represent the exponents of the entire row. This step is accomplished by performing an OR operation on all 8 exponent values, and then detecting the leading 1. It will output a 3b number to the packer and compressor output. The exponent lengths need to be stored as metadata per row. These are stored separately, necessitating two write streams per tensor; both streams are sequential, thus DRAM-friendly. Furthermore, because there are only 3 bits per group of 8 values, a single access to this metadata structure yields metadata for multiple groups. Accesses to this metadata structure will thus be much less frequent than that for the value containers.

To avoid wide crossbars when packing/unpacking, values remain within the confines of their original format bit positions as per the method proposed in Proteus (Judd et al., 2016b). In contrast to Proteus, however, here every row uses a different bitlength, the values are floating-point, the bitlengths vary during runtime and per row, and we target training. Each packer, shown in Figure 11, takes a single FP32 number masks out unused exponent and mantissa bits, and rotates the remaining bits to a position to fill in the



Figure 9: Compressor.



Figure 10: Decompressor.

output row. The mask is created based on the exp_width and man_width inputs. The rotation counter register provides the rotation count which is updated to (exp_width+man_width) every cycle. The (L,R) register pair is used to tightly pack the encoded values into successive rows. They are needed since a value may now be split across two memory rows. This arrangement effectively packs the values belonging to each column tightly within a column of 32b in memory. Since each row is the same total bitlength, the 8 packers operate in tandem filling their respective outputs at exactly the same rate. As a result, the compressor produces $8 \times 32b$ at a time. The rate at which the outputs are produced depends on the compression rate achieved, the higher the compression, the lower the rate.

Decompressor: As Figure 10 shows, the decompressor mirrors the compressor. The inputs to the unit are a 3b exp_width, a 5b man_width, and a $8 \times 32b$ compressed data input. Since the data is compressed, a single row of $8 \times 32b$ will typically contain data from more than one original uncompressed row of FP32 numbers. The compressed data values are packed into 8 virtual columns withing each row. Accordingly, each of the 8 virtual columns of 32b is fed into a dedicated unpacker.

Each unpacker, shown in Figure 12, has a wide 64b register that is internally divided into L and R registers of 32b. They are used in a similar fashion as the corresponding registers of the packer unit. At any point in time, one of the registers is used to accept a new row of 32b packed



Figure 11: Packer.



Figure 12: Unpacker.

Table 3:	Hardware	Area C	Overhead.
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module	area per unit (um^2)	unit number	total area (mm^2)
compressor	31575.60	16	0.505
decompressor	37133.28	16	0.594
accelerator	38533.68	8000	308.27

data whereas the other contains whichever bits from the previous row of 32b have not been used yet. The combineand-shift will combine the input data and previous data in the register then shift to the left. The number of shifted bits is determined by the exponent and mantissa lengths of this row. The 32b data on the left of the register are taken out and shifted to the right (zero extending the exponent). Finally, the unpacker reinserts the mantissa bits that were trimmed during compression. Since each row of data uses the same total bitlength, the unpackers operate in tandem consuming data at the same rate. The net effect is that external memory see wide accesses on both sides.

B HARDWARE EVALUATION METHODOLOGY

Best practices for the evaluation of custom hardware architectures necessitate exploration and validation first via analytical modelling or via cycle-accurate simulation. Since training these networks takes several days on actual hardware, cycle-accurate simulation of the full process is impractical. To estimate performance and energy, we use the best practice approach by analytically modelling the time and energy used per layer per pass of a baseline accelerator. To do so, we use traffic and compute counts collected during the aforementioned full training runs. We record these counts each time a layer is invoked using PyTorch hooks. We model time and energy for memory accesses via DRAMSIM3 (Li et al., 2020). For modeling on-chip structures we use CACTI (HewlettPackard) for the buffers and layout measurements for the compute units and the Gecko compressors/decompressors. We use a commercial 65nm process to model the processing units and Gecko hardware. We implement the units in Verilog and perform synthesis via the Synopsys Design Compiler and *layout* using Cadence Innovus with a target frequency of 500MHz. Synthesis uses Synopsys' commercial Building Block IP library for the target tech node. We estimate power via Innovus using traces over a representative input sample to model properly signal activity. We used nominal operating conditions to model power and latency. There are two Gecko compressor/decompressor units per channel.

Due to the complexity and time cost of cycle-accurate hardware simulation, we have opted for an estimated time and energy consumption analytical model based on the proposed hardware description and the compressor-decompressor architecture. To compute the analytical model, we first analyze the network and retrieve its structure (layer input and output sizes, kernel sizes for convolutional layers, stride, bias and padding). We then calculate the compute operations that will happen for the general batch size (N) in both the forward and backward pass, as well as the number of parameters that must be stored in memory for activations, weights and gradients.

To take advantage of data reuse where possible we perform the forward pass in a layer-first order per batch. This allows us to read the weights per layer only once per batch. For the backward pass, we utilize the on-chip buffers for mini-batching with a layer-first order over a mini-batch of samples. Mini-batching reduces overall traffic by processing as many samples as possible in a layer-first order avoiding either having to spill gradients or reading and writing weights per sample per layer. The number of samples that can fit in a mini-batch depends on the layer dimensions and the size of the on-chip buffer.

Both SFP_Q and SFP_{BW} sample bitlengths per batch to a log file for both mantissas and exponents. These bitlengths are used to compute the number of mini-batches that can fit at every training step per layer on chip. Based on the number of sampled mini-batches (K) we compute the memory footprint generated on the forward pass for each method. After this, we calculate the footprint that stays on-chip and can be loaded from on-chip for the backward pass, and the footprint that goes to off-chip and has to be loaded to on-chip again for it. Based on these memory accesses, we use DRAMsim to simulate the number of compute-cycles that take the memory accesses to finish and we use the maximum cycles between compute and memory as the time constraint to calculate total computation time in the proposed hardware.

To calculate energy consumption and efficiency, we use the information gathered in terms of on-chip memory access cycles, off-chip memory access cycles and compute cycles. We estimate energy consumption for all components including the compressors and decompressors. We use the following equations to estimate energy consumption for our methods (all symbols are defined in Table 4):

$$E_{forward} = E_{compute fwd} + E_{offchip in act mem} + E_{offchip wgt mem} + E_{offchip out act mem} + E_{onchip in act mem} + E_{onchip in act mem} + E_{conchip wgt mem} + E_{onchip out act mem} + E_{read ops mem} + E_{decomp act} + E_{decomp wgt} + E_{comp act}$$

$$(17)$$

$$E_{backward} = E_{compute bck} + E_{offchip in act mem} + E_{offchip wgt mem} + E_{onchip in act mem} + E_{onchip wgt mem} + E_{read ops mem} + E_{decomp act} + E_{decomp wat}$$
(18)

where,

$$E_{offchip in act mem} = \frac{MemCh \times P_{DRAM}}{Freq_{compute}} \times Cycles_{offchip in act}$$
(19)

$$E_{offchip wgt mem} = \frac{MemCh \times P_{DRAM}}{Freq_{compute}} \times$$

$$(Cycles_{offchip wgt} + Cycles_{offchip wgt grad})$$
(20)

$$E_{offchip out act mem} = \frac{MemCh \times P_{DRAM}}{Freq_{compute}} \times Cycles_{offchip out act}$$
(21)

$$E_{onchip in act mem} = Cycles_{onchip in act write} \times P_{onchip write}$$
(22)

$$E_{onchip wgt mem} = Cycles_{onchip wgt read} \times P_{onchip read}$$
(23)

 $E_{onchip out act mem} = Cycles_{onchip out act read} \times P_{onchip read} + Cycles_{onchip out act write} \times P_{onchip write}$ (24)

$$E_{decomp} = P_{decomp (comp \, ratio)} \times \frac{Cycles_{comp \, to \, decomp}}{Freq_{compute}}$$
(25)
$$E_{comp} = P_{comp (comp \, ratio)} \times \frac{Cycles_{decomp \, to \, comp}}{Freq_{compute}}$$
(26)

$$E_{decomp act} = E_{decomp act(comp ratio)}$$
(27)

$$E_{decomp wgt} = E_{decomp wgt(comp ratio)}$$
(28)

$$E_{comp \, act} = E_{comp \, act(comp \, ratio)} \tag{29}$$

Table 4: Symbols definition table.

Symbol	Definition		
	Energy consumption of the compute module for		
$E_{compute fwd}$	the entirety of the computations in the forward pass		
$E_{\it compute bck}$	Energy consumption of the compute module for		
	the entirety of the computations in the backward pass		
$E_{\it off} chip$ in act mem	Energy consumption of the offchip memory transfers		
	for the network input activations		
$E_{offchipwgtmem}$	Energy consumption of the offchip memory transfers		
	for the network weights		
$E_{\it off}$ chip out act mem	Energy consumption of the offchip memory transfers		
	for the network output activations		
E	Energy consumption of the onchip memory transfers		
$E_{onchip in act mem}$	for the network input activations		
$E_{onchip wgt mem}$	Energy consumption of the onchip memory transfers		
	for the network weights		
$E_{onchipout act mem}$	Energy consumption of the onchip memory transfers		
	for the network output activations		
E read ops mem	Energy consumption of loading		
	operations from memory		
	Energy consumption of decompressing		
$E_{decompact}$	activations in the decompressor		
$E_{decomp wgt}$	Energy consumption of decompressing		
	weights in the decompressor		
$E_{\it compact}$	Energy consumption of compressing		
- comp act	activations in the compressor		
$P_{decomp(compratio)}$	Power consumption by the decompressor when loading data		
	from offchip memory at a specific compression ratio (see Table 5)		
$P_{comp(compratio)}$	Power consumption by the compressor when writing data		
	to offchip memory at a specific compression ratio (see Table 5)		
MemCh	Number of available memory channels		
P_{DRAM}	Power consumption of offchip DRAM		
Freq _{compute}	Clock frequency of the hardware accelerator		
Cycles offchip in act	Compute cycles taken to read input activations from offchip memory		
Cycles offchip wgt	Compute cycles taken to read weights from offchip memory		
Cycles offchip wgt grad	Compute cycles taken to read weight gradients from offchip memory		
Cycles offchip out act	Compute cycles taken to read output activations from offchip memory		
Cycles _{onchip} in act write	Compute cycles taken to read input activations from onchip memory		
Cycles onchip wgt read	Compute cycles taken to read weights from onchip memory		
Cycles onchip out act read	Compute cycles taken to read output activations from onchip memory		
Cycles _{onchip} out _{act} write	Compute cycles taken to write output activations to onchip memory		
Ponchip write	Power consumption of a word write to onchip memory		
Ponchipread	Power consumption of a word read from onchip memory		
Cycles _{comptodecomp}	Compute cycles taken to decompress compressed data		
$Cycles_{decomp to comp}$	Compute cycles taken to compress data		

Compressor power (mW) Decompressor power (mW) Compression ratio 0.143 - 0.263 10.87 13.84 0.264 - 0.388 12.18 14.72 0.389 - 0.513 12.65 15.97 0.514 - 0.638 13.44 15.76 14.98 0.639 - 0.763 15.42

Table 5: P() terms: Power consumption as a function compression ratio.